

# One Channel H-Bridge Motor Driver AM1016A

## ● Features and Benefits

- Wide Supply Voltage Range (2.0V~6.8V)
- 0.9A Continuous Current, 2A Peak Current
- Low Standby Mode Current (Typ=0.01μA)
- Low Quiescent Operation Current
- Low MOSFETs On-resistance 0.6Ω@Io=0.2A;  
0.65Ω@Io=0.6A
- Provide Four Operation Modes:  
Forward/Reverse/Stop/Brake
- Thermal Shutdown Protection
- SOT23-6 Package for Small Size PCB Layout
- Pb-Free and Halogen-Free Green Product

## ● Applications

- Toys
- Small Appliances
- Robotics
- Consumer Products

## ● Description

The AM1016A is one channel H-Bridge driver. It provides integrated motor driver solution for toys, robotics, consumer products and other low voltage or battery-powered motion control applications.

The AM1016A can supply up to 0.9 A continuous current and 2.0 A peak current. There is internal shutdown function for over-temperature protection.

The small package, reducing 20% package size compared with SOP8 size, have a miniaturizing advantage even more.

Package material is Pb-Free and Halogen-Free (Green) for the purpose of environmental protection and for sustainable development of the Earth.

## ● Ordering Information

Orderable Part Number	Package
AM1016A	SOT23-6

● **Absolute Maximum Ratings ( $T_A = 25^\circ\text{C}$ )**

Parameter	Symbol	Limits	Unit
Power Supply voltage	VCC	7.0	V
Output Continuous Current	$I_{O\_CONT}$	0.9 *	A
Output Peak Current	$I_{O\_Peak}$	2.0	A
Operate Temperature Range	$T_{opr}$	-40~+85	$^\circ\text{C}$
Storage Temperature Range	$T_{stg}$	-40~+150	$^\circ\text{C}$

\*Based on 20x20 mm<sup>2</sup> FR4 PCB (1 oz.) at single side PCB

● **Recommended operating conditions ( $T_A = 25^\circ\text{C}$ )**

(Set the power supply voltage taking allowable dissipation into considering)

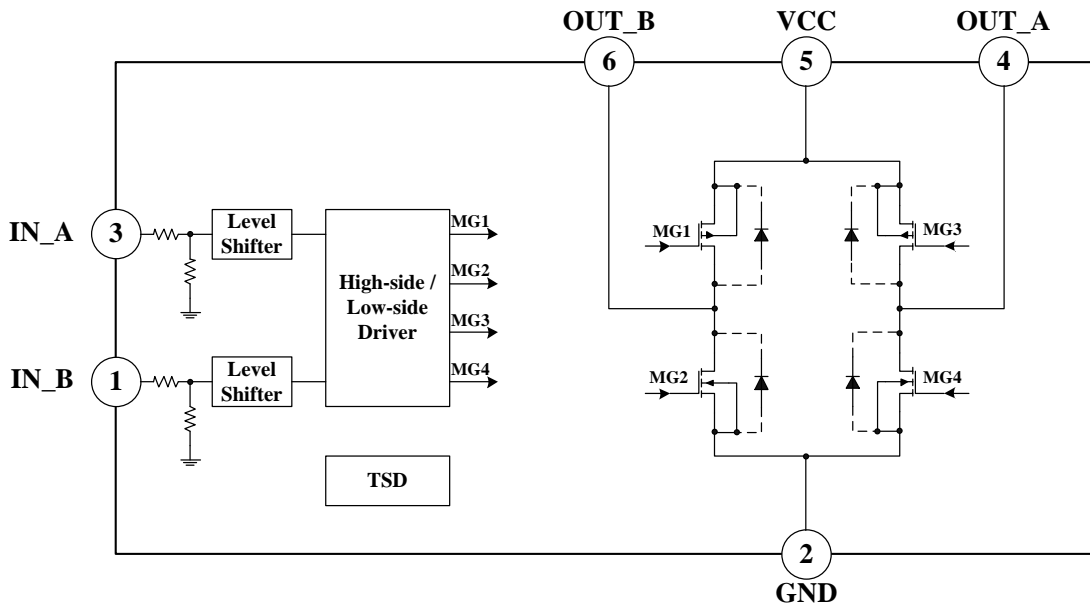
Parameter	Symbol	Min	Typ	Max	Unit
Power Supply Voltage	VCC	2.0		6.8	V
IN_A and IN_B	$V_{IN\_X}$	-0.3		Vcc+0.3	V
Output Continuous Current	$I_o$	0		0.9 *	A
Externally Applied PWM Frequency	$F_{IN\_X}$	0.02		65	KHZ

\*Based on 20x20 mm<sup>2</sup> FR4 PCB (1 oz.) at single side PCB

● Electrical Characteristics ( Unless otherwise specified,  $T_A = 25^\circ\text{C}$ ,  $V_{CC}=5\text{V}$ )

Parameter	Symbol	Limit			Unit	Conditions
		Min	Typ	Max		
<b>Power Supplies</b>						
Quiescent Operation Current	$I_{CC}$		65		$\mu\text{A}$	Input signal IN_A/B= L/H or H/L or H/H, No load on OUT_A/B
Standby Current	$I_{STB}$		0.01	1	$\mu\text{A}$	Input signal IN_A/B= L/L, No load on OUT_A/B
<b>PWM Inputs</b>						
Input H Level Voltage	$V_{IN\_XH}$	2.0		$V_{CC}$	V	
Input L Level Voltage	$V_{IN\_XL}$	-0.3		0.7	V	
Input H Level Current	$I_{IN\_X}$		30		$\mu\text{A}$	$V_{CC} = 5\text{V}$ , $V_{IN\_X} = 3\text{V}$
Input Frequency	$F_{IN\_X}$	0.02		65	kHz	
Input Pull-Down Resistance	$R_{IN\_X}$		100		K $\Omega$	
<b>H-Bridge FETs</b>						
On-Resistance	$R_{ds(on)}$		0.6		$\Omega$	$I_o = 200\text{mA}$ Upper and Lower total
On-Resistance	$R_{ds(on)}$		0.65		$\Omega$	$I_o = 600\text{mA}$ Upper and Lower total
<b>TSD Protections</b>						
Thermal Shutdown Protection	$TSD_p$		150		$^\circ\text{C}$	
Thermal Shutdown Release	$TSD_r$		125		$^\circ\text{C}$	

● Block Diagram



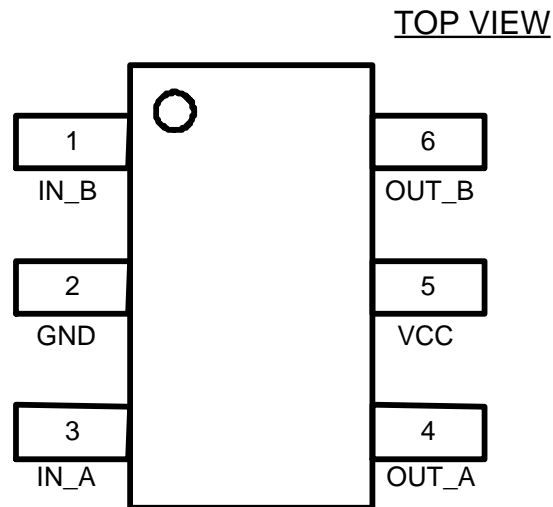
● Input Logic Descriptions

Function truth table

IN_A	IN_B	OUT_A	OUT_B	Mode
L	L	Hi-Z	Hi-Z	Stop
L	H	L	H	Reverse
H	L	H	L	Forward
H	H	L	L	Brake

※Low standby mode current function when IN\_A = IN\_B = Low level

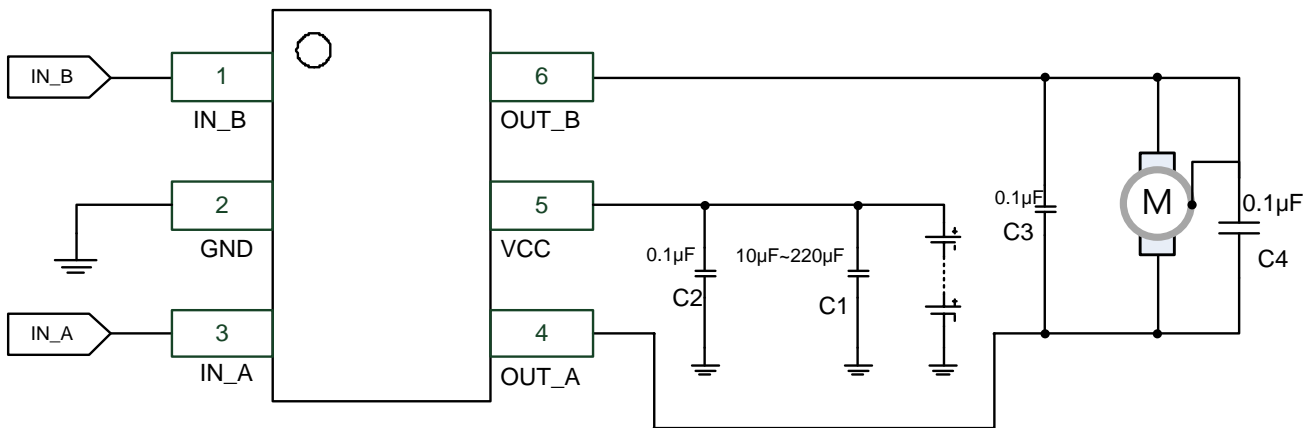
● Pin configuration SOT23-6



● Pin Descriptions

PIN Number	Pin Name	I/O	Description
1	IN_B	I	Input Half Bridge B
2	GND	-	Ground Pin
3	IN_A	I	Input Half Bridge A
4	OUT_A	O	Output Half Bridge A
5	VCC	-	Power Supply
6	OUT_B	O	Output Half Bridge B

## ● Application



## ● Circuit Descriptions

The function descriptions of capacitors on the application circuit:

C1 、 C2: Power supply VCC pin capacitor:

- 1) The capacitor can reduce the power spike when the motor is in motion. To avoid the IC directly damaged by the VCC peak voltage. It also can stabilize the power supply voltage and reduce its ripples.
- 2) The C1 capacitor can compensate power when motor starts running.
- 3) The capacitor value ( $\mu\text{F}$ ) determines the stability of the VCC during motor in motion. If the large voltage power or a heavy loading motor is used, then a larger capacitor would be needed.
- 4) On the PCB configuration, the C1 、 C2 must be mounted as close as possible to VCC pin (PIN5).

C3: The across-output capacitor ; C4: The across-motor capacitor

- 1) The capacitors can reduce the power spike of motor when operating. Therefore, a  $0.1\mu\text{F}$  capacitor is recommended.
- 2) On the PCB configuration, the C3 must be mounted as close as possible to OUT\_A&B (PIN 4&PIN 6).
- 3) The C4 capacitor single-ended can be welded on the motor shell.
- 4) The C3 、 C4 capacitor must be added to the general application.

### ● Operating Mode Descriptions

#### 1) H-Bridge basic operating mode :

##### a) Forward mode

Definition : When  $IN\_A=H$  ,  $IN\_B=L$  , then  $OUT\_A=H$  ,  $OUT\_B=L$

##### b) Reverse mode

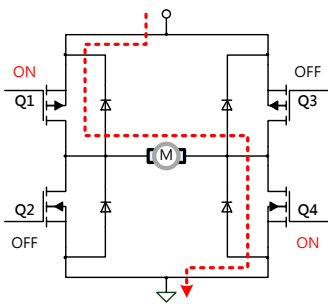
Definition : When  $IN\_A=L$  ,  $IN\_B=H$  , then  $OUT\_A=L$  ,  $OUT\_B=H$

##### c) Stop/Brake mode

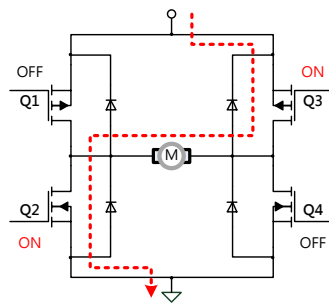
Definition : When  $IN\_A=IN\_B=H$  , then  $OUT\_A=OUT\_B=L$

##### d) Stop mode

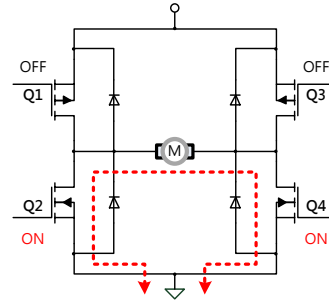
Definition : When  $IN\_A=IN\_B=L$  , then  $OUT\_A=OUT\_B=Hi-Z$



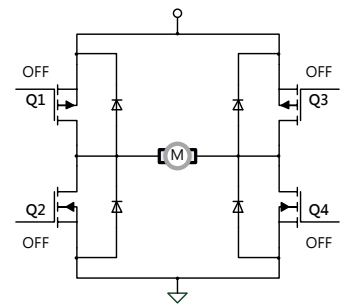
a) Forward mode



b) Reverse mode



c) Brake mode



d) Stop mode

### ● Protection Mechanisms Descriptions

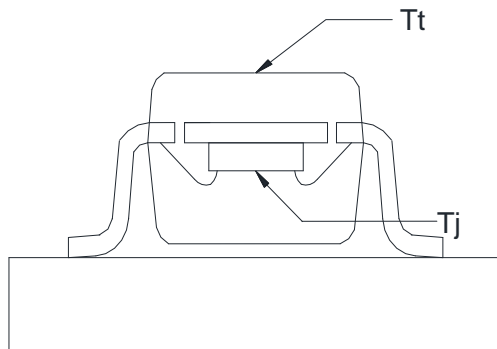
#### 1) Over-temperature protection

If the IC junction temperature exceeds  $150^{\circ}C$  (Typ.), the internal over-temperature protection function will be triggered, and all FETs in the H-bridge are disabled, that will ensure the safety of customers' products. If the IC junction temperature falls to  $125^{\circ}C$  (Typ.), the IC resumes automatically.

● Thermal Information

$\theta_{ja}$	junction-to-ambient thermal resistance	277.78°C/W
$\Psi_{jt}$	junction-to-top characterization parameter	20.6°C/W

- $\theta_{ja}$  is obtained in a simulation on a JEDEC-standard 1s0p board as specified in JESD-51.
- The  $\theta_{ja}$  number listed above gives an estimate of how much temperature rise is expected if the device was mounted on a standard JEDEC board.
- When mounted on the actual PCB, the  $\theta_{ja}$  value of JEDEC board is totally different than the  $\theta_{ja}$  value of actual PCB.
- $\Psi_{jt}$  is extracted from the simulation data to obtain  $\theta_{ja}$  using a procedure described in JESD-51, which estimates the junction temperature of a device in an actual PCB.
- The thermal characterization parameter,  $\Psi_{jt}$ , is proportional to the temperature difference between the top of the package and the junction temperature. Hence, it is useful value for an engineer verifying device temperature in an actual PCB environment as described in JEDEC JESD-51-12.
- When Greek letters are not available,  $\Psi_{jt}$  is written Psi-jt.
- Definition:



$$\Psi_{jt} = (T_j - T_t) / P_d$$

Where :

$\Psi_{jt}$  (Psi-jt) = Junction-to-Top(of the package) °C/W

$T_j$ = Die Junction Temp. °C

$T_t$ = Top of package Temp at center. °C

$P_d$ = Power dissipation. Watts

- Practically, most of the device heat goes into the PCB, there is a very low heat flow through top of the package, so the temperature difference between  $T_j$  and  $T_t$  shall be small, that is any error caused by PCB variation is small.
- This constant represents that  $\Psi_{jt}$  is completely PCB independent and could be used to predict the  $T_j$  in the environment of the actual PCB if  $T_t$  is measured properly.
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## ● How to predict $T_j$ in the environment of the actual PCB

Step 1 : Used the simulated  $\Psi_{jt}$  value listed above.

Step 2 : Measure  $T_t$  value by using

### ➤ Thermocouple Method

We recommend use of a small ~40 gauge(3.15mil diameter) thermocouple. The bead and thermocouples wires should touch the top of the package and be covered with a minimal amount of thermally conductive epoxy. The wires should be heat-insulated to prevent cooling of the bead due to heat loss into wires. This is important towards preventing “too cool”  $T_t$  measurements, which would lead to the calculated  $T_j$  also being too cool.

### ➤ IR Spot Method

An IR Spot method should be utilized only when using a tool with a small enough spot area to acquire the true top center “hot spot”.

Many so-called “small spot size” tools still have a measurement area of 0~100+mils at “zero” distance of the tool from the surface. This spot area is too big for many smaller packages and likely would result in cooler readings than the small thermocouple method.

Consequently, to match between spot area and package surface size is important while measuring  $T_t$  with IR spot method.

Step 3 : calculating power dissipation by

$$P \cong (VCC - |V_{o\_Hi} - V_{o\_Lo}|) \times I_{out} + VCC \times I_{cc}$$

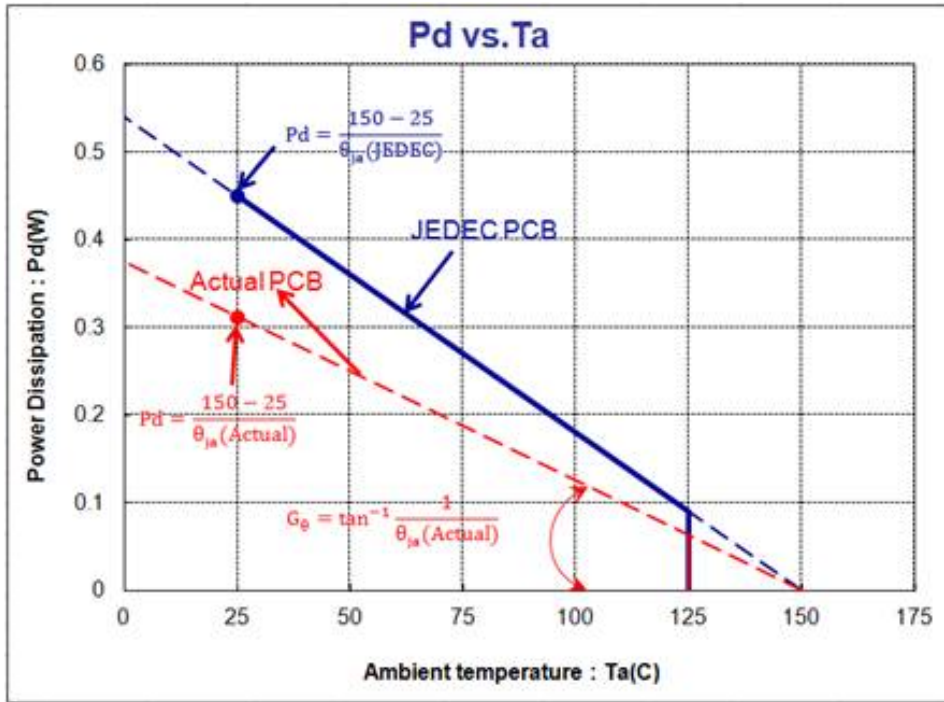
Step 4 : Estimate  $T_j$  value by

$$T_j = \Psi_{jt} \times P + T_t$$

Step 5: Calculated  $\Theta_{ja}$  value of actual PCB by the known  $T_j$

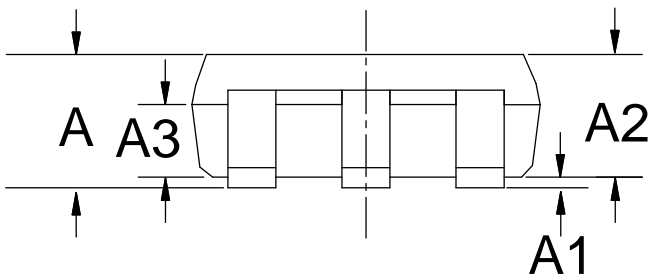
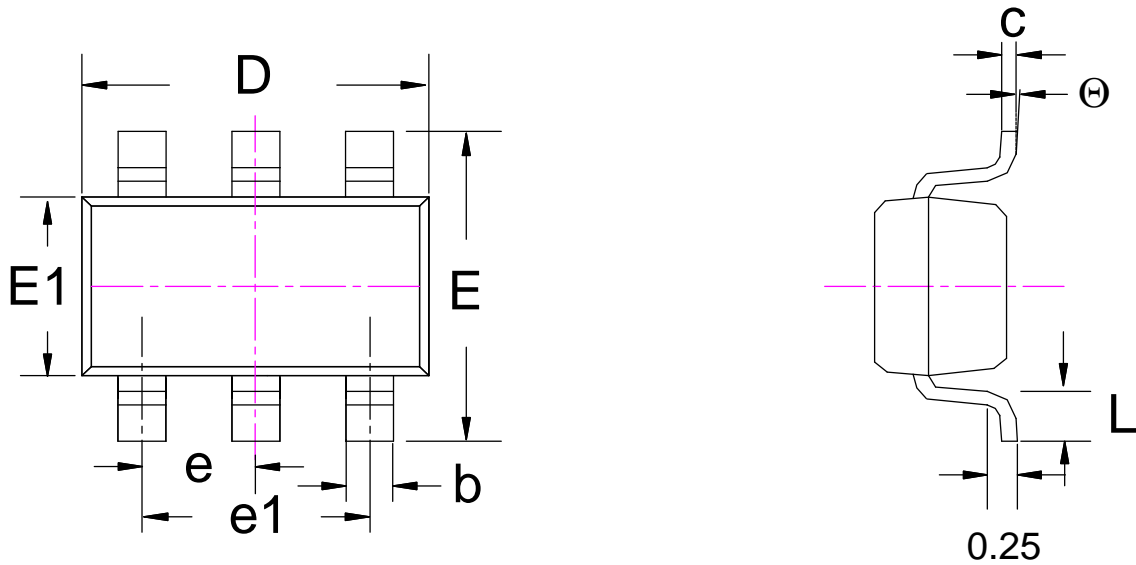
$$\Theta_{ja}(\text{actual}) = (T_j - T_a) / P$$

- Maximum Power Dissipation (de-rating curve) under JEDEC PCB & actual PCB



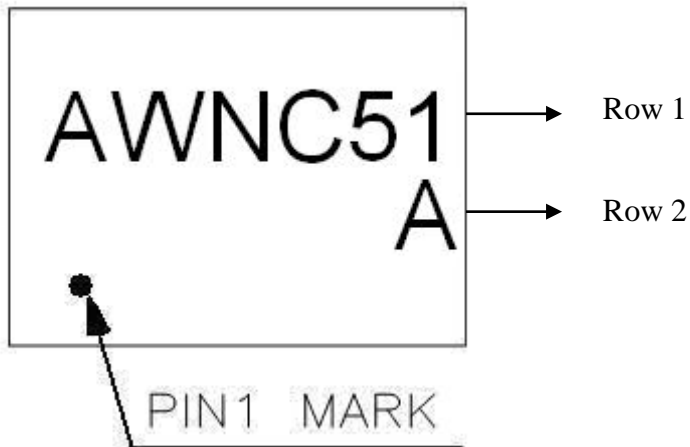
● Packaging outline --- SOT23-6

Unit: mm



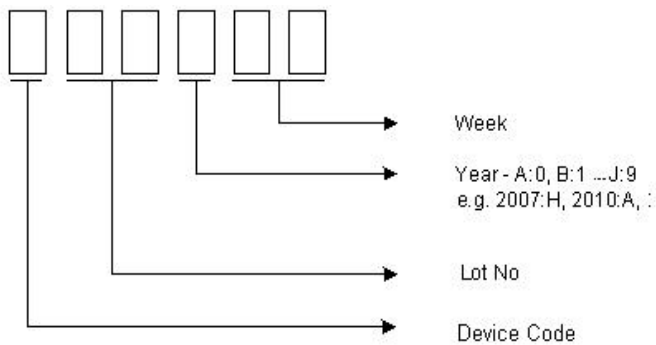
SYMBOL	MILLIMETERS		INCHES	
	Min.	Max.	Min.	Max.
A	--	1.35	--	0.053
A1	0.04	0.15	0.002	0.006
A2	1.00	1.20	0.039	0.047
A3	0.55	0.75	0.022	0.030
b	0.34	0.43	0.013	0.017
c	0.15	0.21	0.006	0.008
D	2.72	3.12	0.107	0.123
E	2.60	3.00	0.102	0.118
E1	1.40	1.80	0.055	0.071
L	0.3	0.6	0.012	0.024
⊖	0	8	0	0.315
e	0.95 BSC		0.037 BSC	
e1	1.90 BSC		0.075 BSC	

● **Marking Identification**



Row 1

Device Code, Lot number & Date



Row 2

Customer specific difference code

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